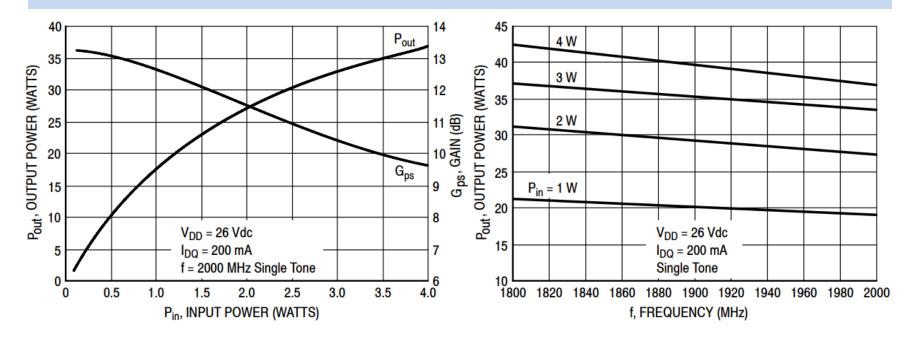
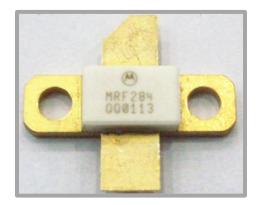
MICROSTRIP PA DESIGN

PART – I (MRF284 on 1296)

MRF284 LDMOS TRANSISTOR



- THE MRF284 IS A Si LDMOS POWER TRANSISTOR
- > 30W POWER OUTPUT AT Vdd = +26V
- MOTOROLA / NXP "OBSOLETE"
- EBAY AVAILABLE < \$20



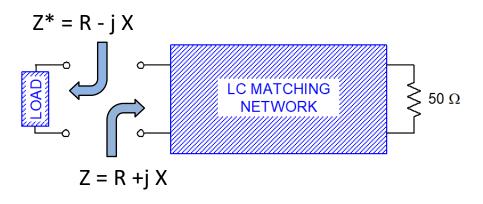
MRF284 TARGET IMPEDANCES

• OPTIMUM EXTERNAL IMPEDANCES at 1296 MHz :

 $Z_{\text{S OPT}} = 2.0 + j 2.5 \Omega$ $Z_{\text{L OPT}} = 3.0 + j 2.5 \Omega$

- OPTIMUM MATCHING NETWORKS WILL PROVIDE A MATCH TO LOADS THAT ARE COMPLEX CONJUGATE OF THE OPTIMUMS
- SO, DESIGN MATCHING NETWORKS TERMINATED WITH:

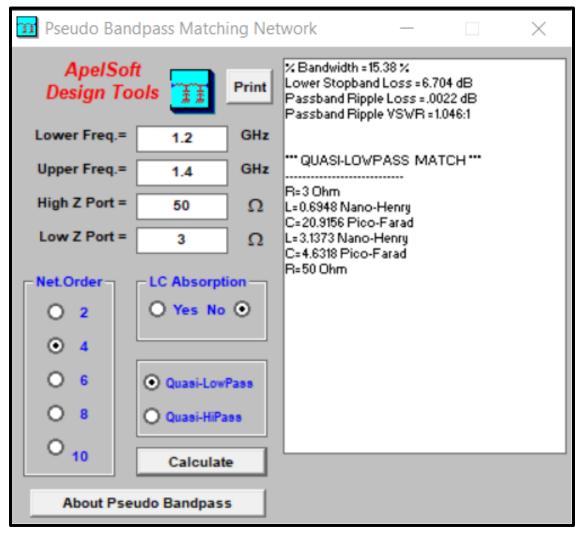
 $Z^*_{S OPT}$ = 2.0 -j 2.5 Ω $Z^*_{L OPT}$ = 3.0 -j 2.5 Ω



INITIAL LUMPED LC DESIGN

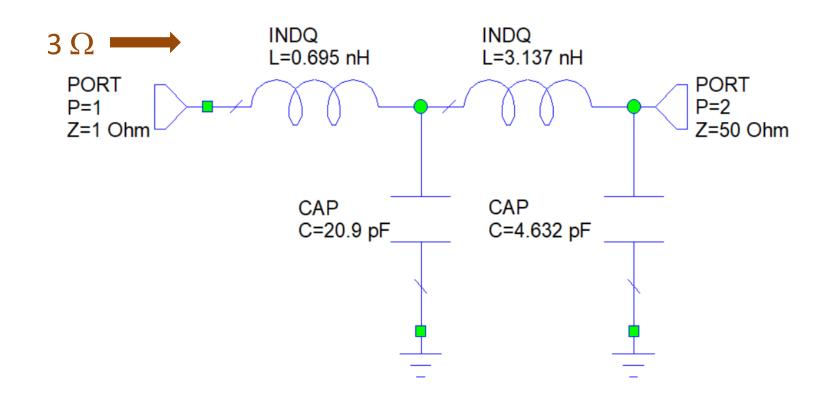
- THE FIRST STEP IN THE DESIGN IS A LUMPED LC MATCH
- THE OUTPUT MATCH WILL BE USED TO ILLUSTRATE THIS
- $Z_{L \text{ OPT}} = 3.0 + j 2.5 \Omega$
- WE WISH TO USE LOWPASS ELEMENTS (L SERIES & C SHUNT)
 - DESIGN A 4th ORDER PSEUDO BP MATCH TO 3 Ω
 - ADD 0.31 nH FOR +j 2.5 Ω

INITIAL 3 OHM MATCH

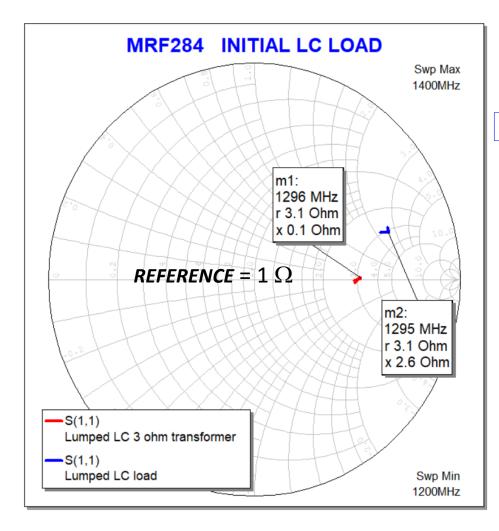


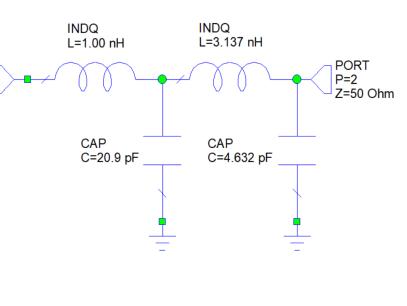
http://k5tra.net/TechFiles/PseudoBPmatch.exe

INITIAL 3 OHM MATCH



LUMPED LC MRF284 OUTPUT MATCH



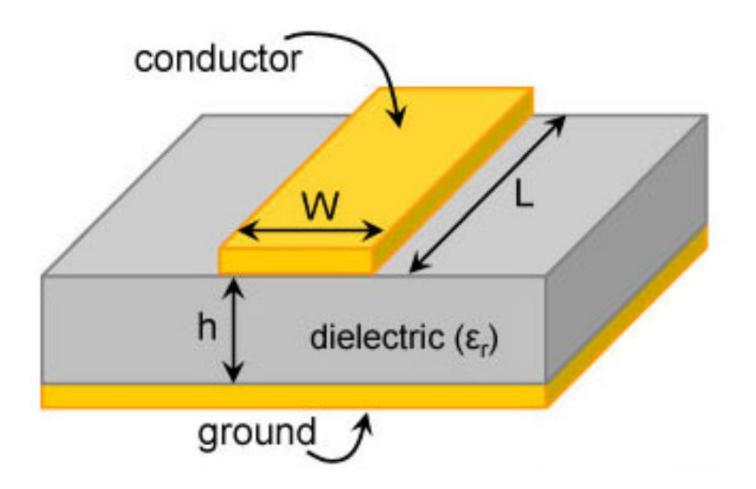


- ADD 0.31 nH to 0.695 nH
- TOTAL L SERIES = 1.00 nH

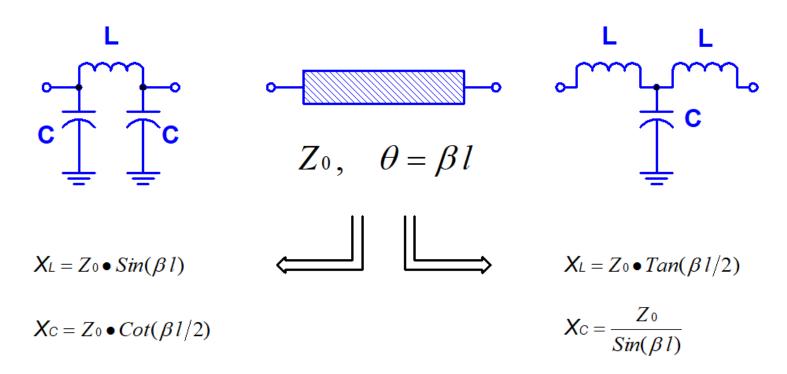
LUMPED DESIGN ON A CIRCUIT BOARD

- THERE REALLY ARE NO LUMPED ELEMENTS
 - CHIP CAPACITORS HAVE SERIES INDUCTANCE (and LOSS)
 - CHIP INDUCTORS HAVE DISTRUBUTED CAPACITNCE (SELF RESONANCE)
 - CHIP RESISTORS HAVE SERIES INDUCTANCE AND SHUNT CAPACITANCE
- CIRCUIT BOARD TRACES OVER BACKSIDE GROUND ARE MICROSTRIP TRANSMISSION LINES.
- MICROSTRIP LINES ARE QUASI-TEM; SO THEY CAN BE REPRESENTED AS A SERIES OF INCREMENTAL SERIES L AND SHUNT C ELEMENTS.
- $Z_0 = \sqrt{\frac{L}{c}}$, WHERE L AND C ARE INCREMENTAL (PER UNIT LENGTH)
- THIN TRACES:
 - HIGH Zo TRACES HAVE HIGH L/C
 - USED FOR PRINTED INDUCTORS
- WIDE TRACES:
 - LOW Zo TRACES HAVE MORE C
 - PROVIDE SHUNT C (AND SOME SERIES L)

MICROSTRIP TRANSMISSION LINES

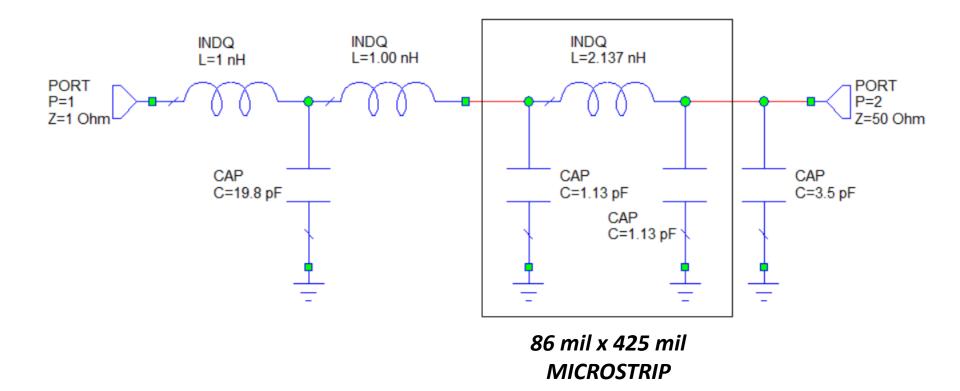


LC REPRESENTATION OF A SHORT LINE

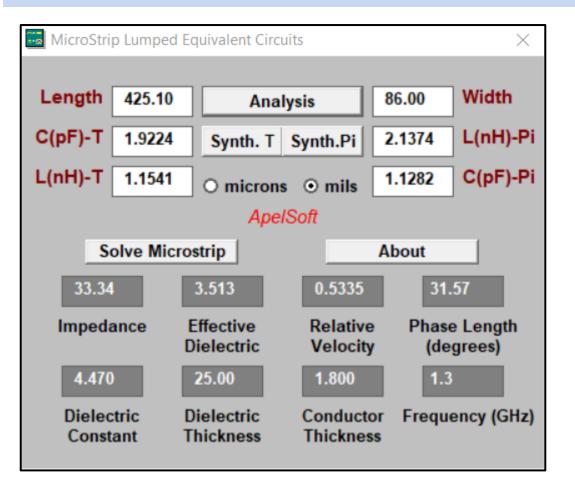


- + HIGH Zo LINES USE THE π CIRCUIT FOR INDUCTOR REPRESENTATION
- THE END CAPACITANCES ARE SMALL WITH HIGH ZO LINES
- LOW ZO LINES USE THE T CIRCUIT FOR CAPACITOR REPRESENTATION
- THE END INDUCTANCES ARE SMALL WITH LOW ZO LINES

PARTITIONED LC OUTPUT MATCH



LUMPED EQUIVALENT CALCULATOR

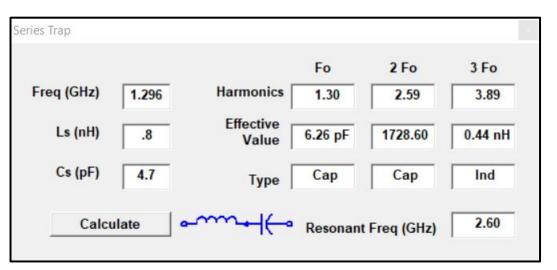


http://k5tra.net/TechFiles/LumpEquiv.exe

- MICROSTRIP ANALYSIS AND SYNTHESIS
- π and T EQUIVALENT CIRCUIT CALCULATION
- SELECTABLE UNITS: MILS OR MICRONS

SMD CHIP SERIES INDUCTANCE

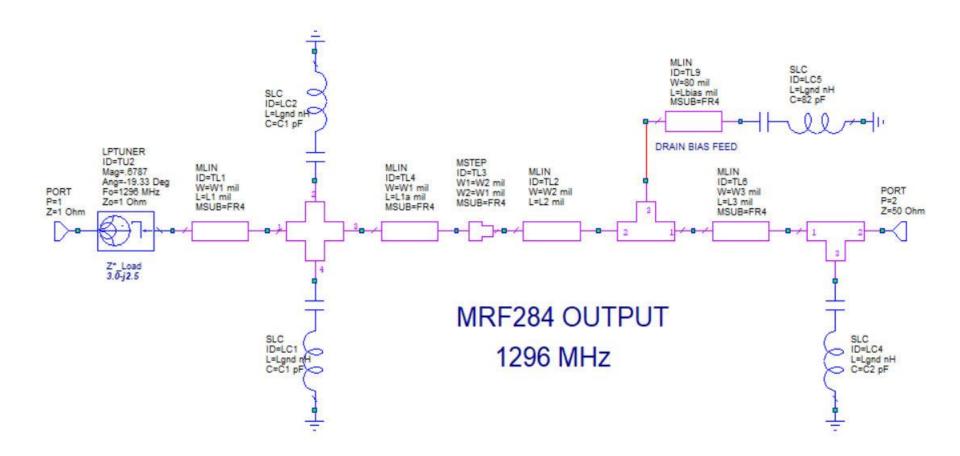
SMD + PTH SERIES INDUCTANCE		
CHIP PKG	DIM (mil ²)	L(nH)
0402	40x20	0.59
0603	60x30	0.77
0805	80x50	0.84
ATC 100A	55x55	0.55
ATC 100B	110x110	0.77



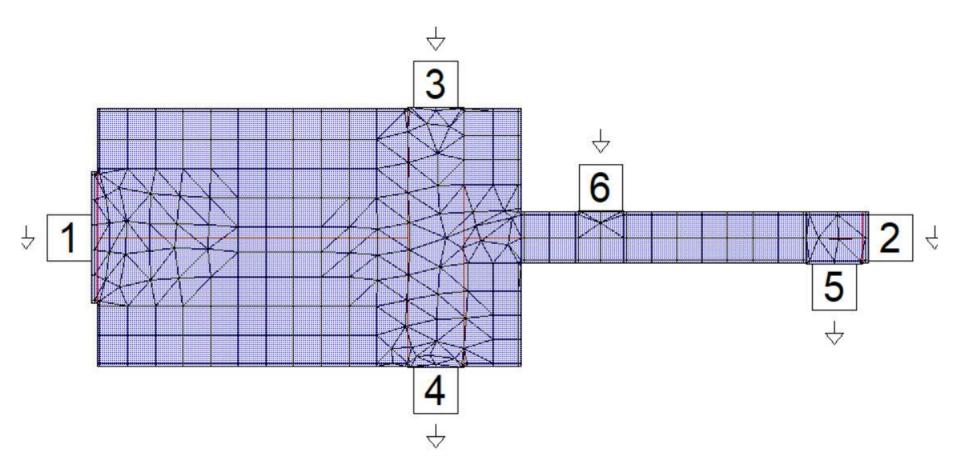
http://k5tra.net/TechFiles/SeriesTrap.exe

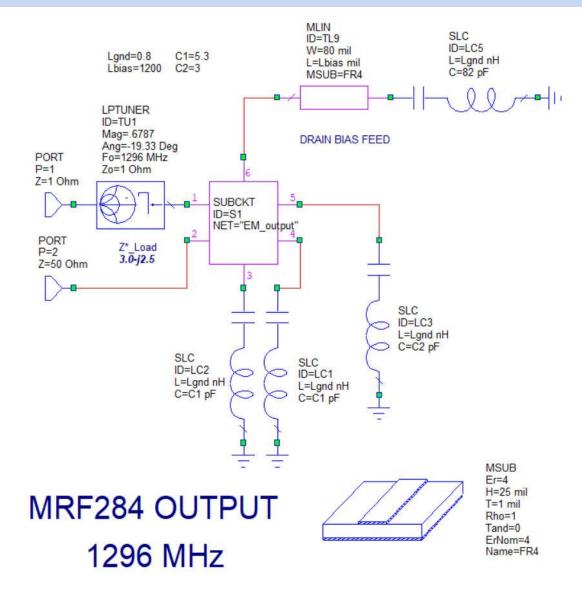
- AN 0805 4.7 pF CAP ALSO HAS 0.8 nH SERIES INDUCTANCE
- SERIES RESONANT FREQUENCY IS 2.59 GHz (2nd HARMONIC SHORT)
- EFFECTIVE CAPACITANCE IS 6.26 pF AT 1296 MHz
- EFFECTIVE CAPACITANCE IS 1720 pF AT 2nd HARMONIC OF 1296 MHz
- EFFECTIVE INDUCTANCE IS 0.44 nH AT 3rd HARMONIC OF 1296 MHz

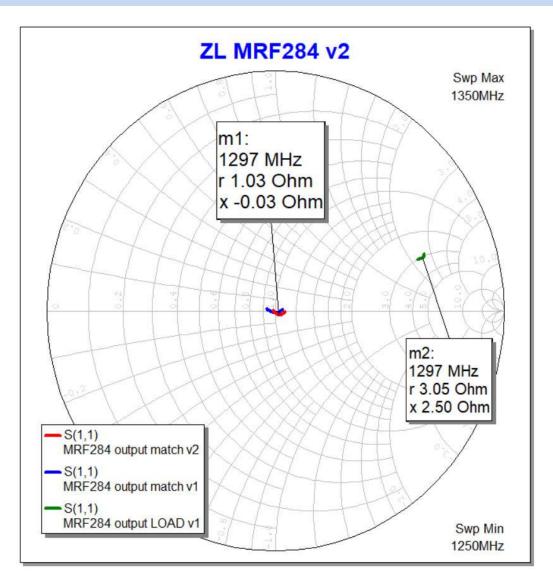
MICROSTRIP DESCRIPTION – MRF286 OUTPUT

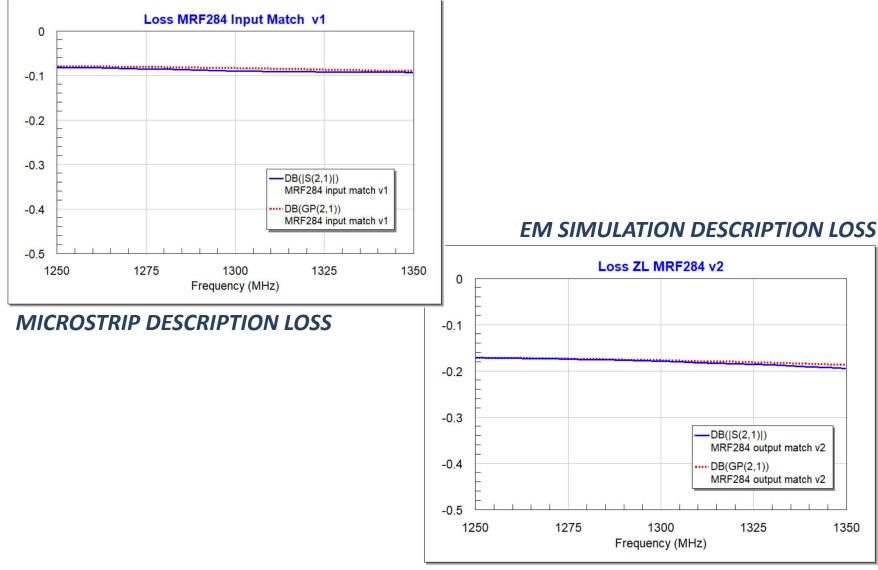


EM SIMULATION GEOMETRY – MRF286 OUTPUT

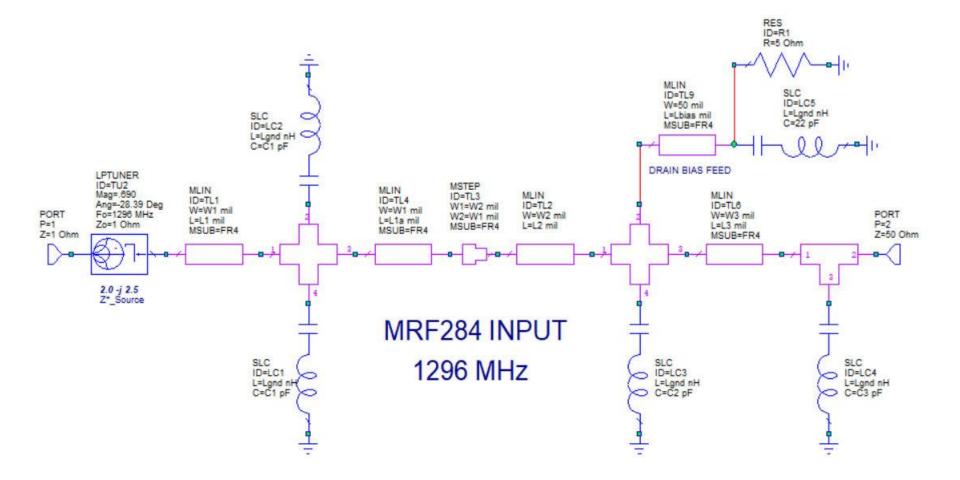




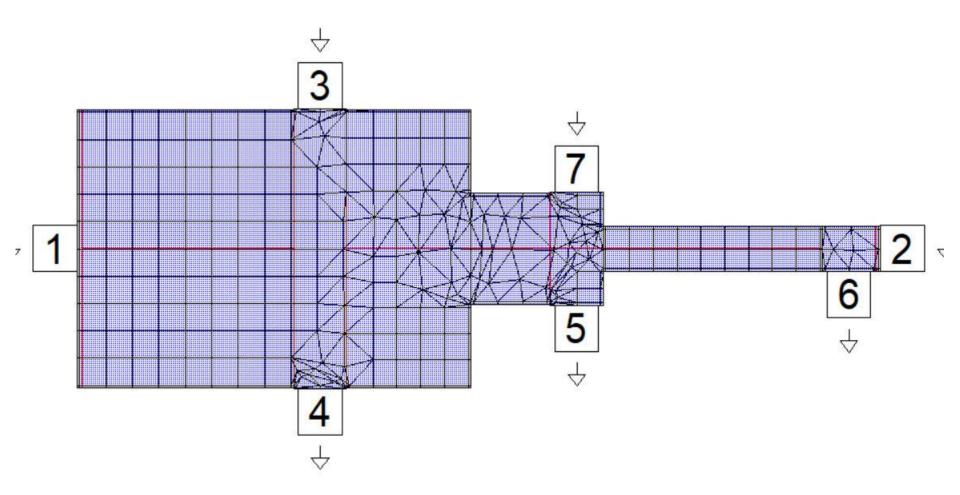


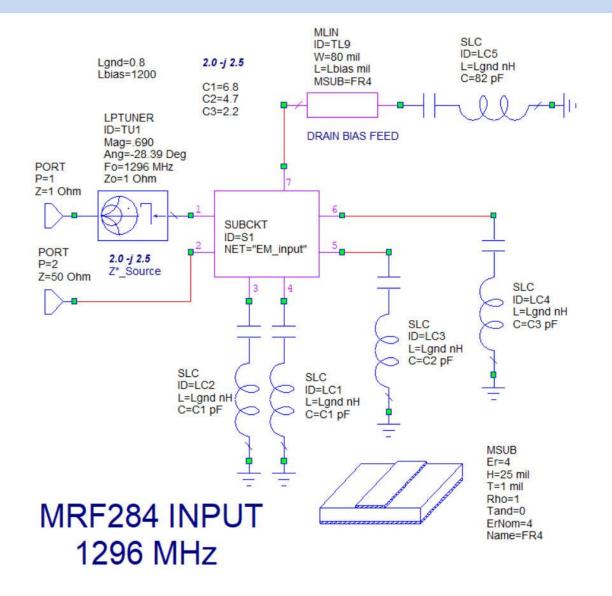


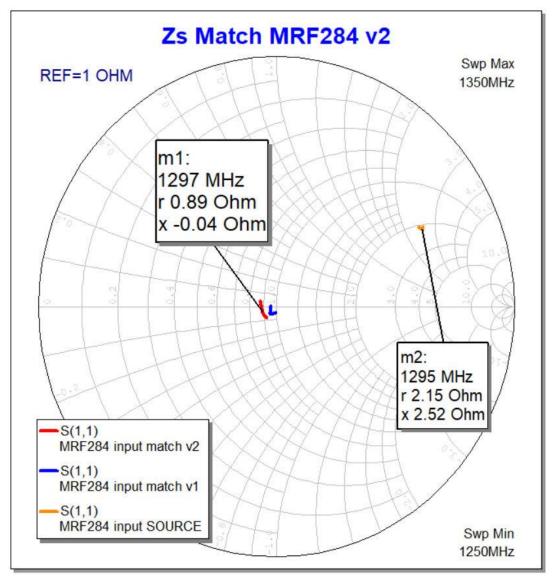
MICROSTRIP DESCRIPTION – MRF286 INPUT

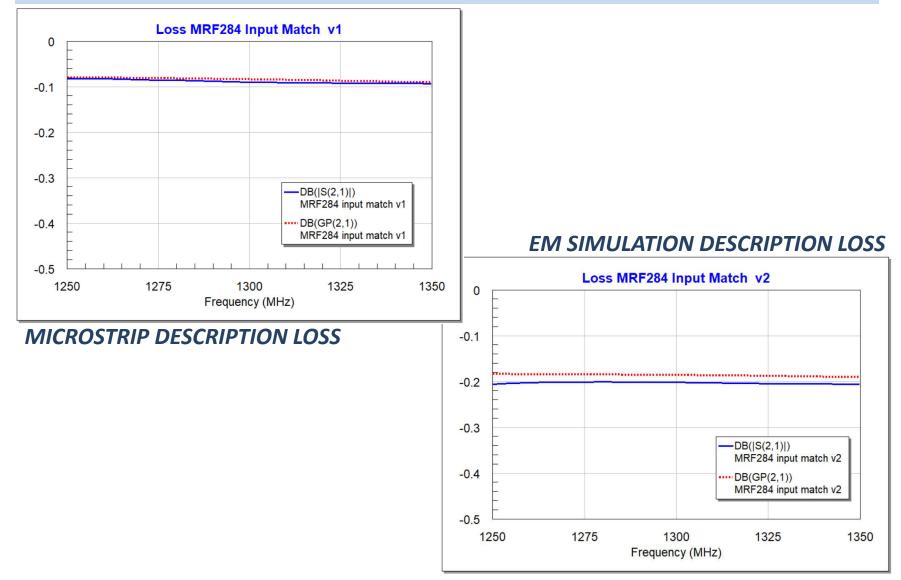


EM SIMULATION GEOMETRY – MRF286 INPUT

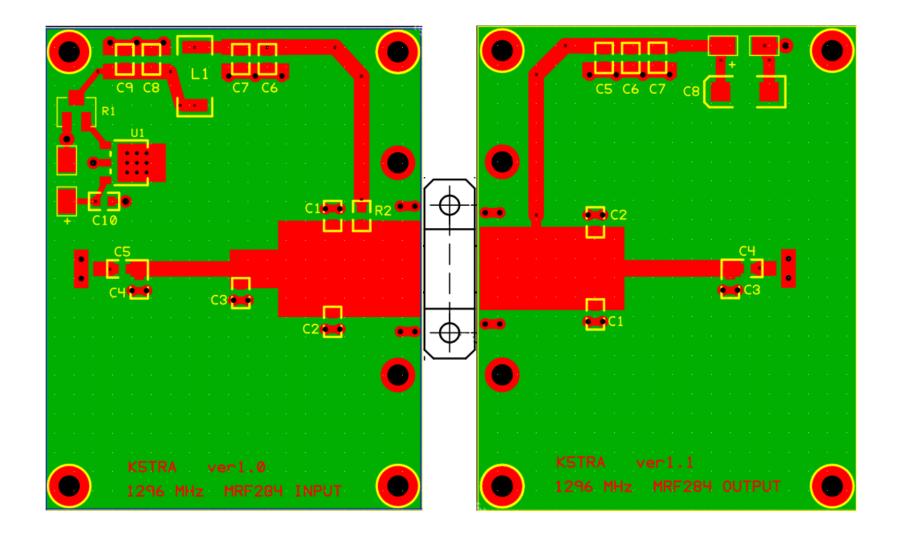








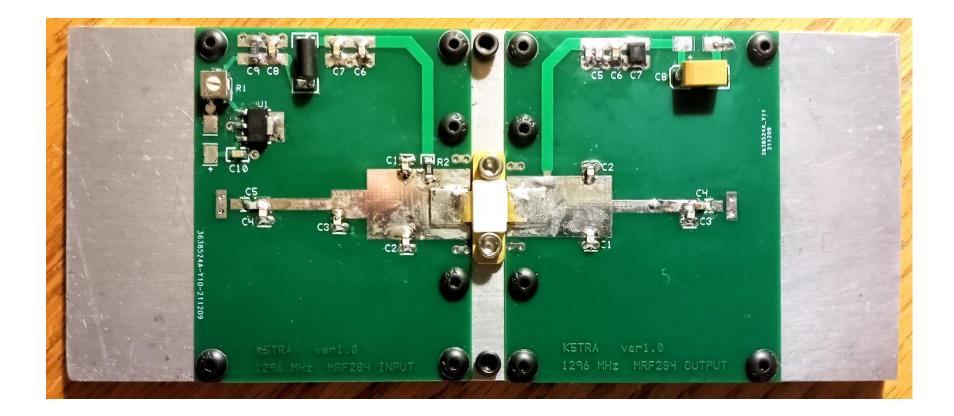
1296 MHz MRF284 30W PA



1296 MHz MRF284 30W PA



1296 MHz MRF284 30W PA



SOFTWARE TOOLS

• SMITH:

http://www.fritz.dellsperger.net/smith.html

- PSEUDO BP MATCH: http://k5tra.net/TechFiles/PseudoBPmatch.exe
- MICROSTRIP LUMPED EQUIVALENT: <u>http://k5tra.net/TechFiles/LumpEquiv.exe</u>
- SERIES TRAP:

http://k5tra.net/TechFiles/SeriesTrap.exe

• SHUNT TRAP:

http://k5tra.net/TechFiles/ShuntTrap.exe